

# Using of a New Bond Graph Technology for the Study of a High Frequency Low Pass Filter

Hichem Taghouti<sup>1</sup>, Sameh Khmailia<sup>2</sup>, Riadh Mehouchi<sup>3</sup> and Abdelkader Mami<sup>4</sup>

<sup>1</sup> Department of Electrical Engineering, National Engineering School of Tunis, Tunisia  
*hichem.taghouti@live.fr*

<sup>2</sup> Department of Electrical Engineering, National Engineering School of Tunis, Tunisia  
*samehkhmailia@gmail.com*

<sup>3</sup> Department of Electrical Engineering, National Engineering School of Tunis, Tunisia  
*rmehouchi.riadh@gmail.com*

<sup>4</sup> Laboratory of High Frequency Circuits, Department of Physics, Sciences Faculty of Tunis,  
2092 El Manar Tunis Tunisia  
*mami.abdelkader@planet.tn*

## Abstract

Due to the frequent uses of electronic filters in radio frequency and telecommunication domains, many researchers have been interesting in its design and modeling for several years.

We will present in this study a new mathematical approach that describes the steps to conceive a high frequency low pass filter by reactive elements then by microstrip line. The scattering bond graph method is applied to determine an explicit model of filter that describes the phenomenon of energy exchange between different elements of system.

**Keywords:** *Low Pass Filter, Scattering Bond Graph, Reflexion and Transmission Coefficients, Modeling and Simulation.*

## 1. Introduction

Electronic filters are electronic circuits which execute signal processing functions, specifically to remove the unwanted frequency components from the signal, to enhance the wanted ones, or both. Filter circuits may be divided into passive and active types. They are widely used in military applications such as radars and satellites and in telecommunication domains. Depending on the characteristics and the conditions of the application, a filter can be designed by reactive elements or by micro-strip line.

In this paper we will present a complete study of high frequency low pass filter. Firstly, we will develop a mathematical approach to determine the order and the normalized elements of filter from its gain curve.

Thereafter, we will deduce the mounting of a filter that is composed by inductances and capacitances components. Then, we will determine a second mounting of filter composed by microstrip lines.

Finally, we will use a scattering bond graph method to analyze the filter.

## 2. Synthesis and calculation of filter order and normalized elements

The synthesis of filter will be summarized by an algorithm that is based on three main steps:

- Determination of inductance and capacitance values by Chebyshev method.
- Determination of the dimensions of the planar filter
- Modeling of filter by scattering bond graph method

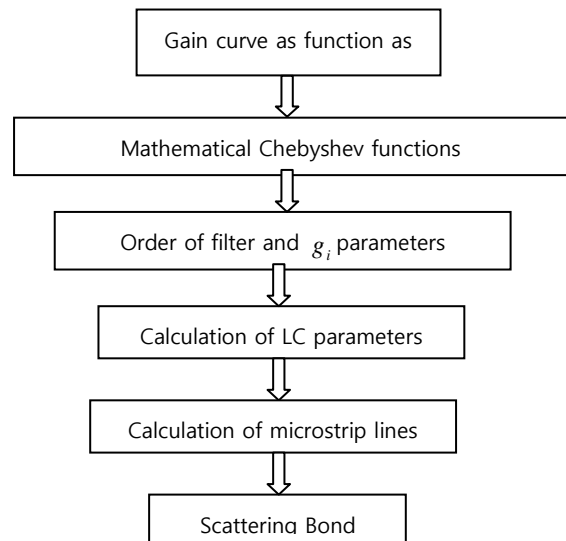


Fig.1 The synthesis algorithm of the filter

The general structure of low pass LC filter [1] is given by the following figure:



Fig.2 General Structure of low pass filter

$R_e = 50\Omega$

The gain curve as function as frequency of the filter that we will design and model is given by figure3:

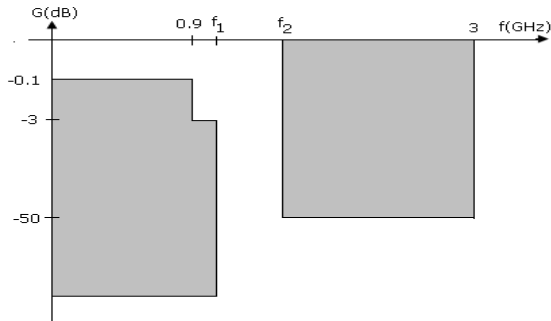


Fig.3 gain curve as function as frequency with a cut-off frequency  $f_1=1\text{GHz}$  in 3 dB and an Attenuation greater than 50 dB from  $f_2=1.5\text{GHz}$  to 3GHz.

From the characteristics given by the gain curve and according to the Chebyshev method, we deduce the order n of filter.

$$n \geq \frac{\text{Argch} \sqrt{\frac{10^{\frac{A_{\min}}{10}} - 1}{10^{\frac{A_{\max}}{10}} - 1}}}{\text{Argch}(w_s)} \quad \text{with} \quad (1)$$

$$\begin{cases} w_s = \frac{f_2}{f_1} \\ A_{\min} = 50\text{dB} \\ A_{\max} = 0.1\text{dB} \end{cases}$$

We obtain  $n=9$ , so the filter is constituted of 9 elements.  
 The normalized parameters  $g_i$  are calculated by the equations given bellow.

$$g_1 = \frac{2}{\eta} \sin\left(\frac{\pi}{2n}\right) \quad (2)$$

$$g_k * g_{k+1} = \frac{4 * \sin\left(\frac{2k-1}{2n} \pi\right) * \sin\left(\frac{2k+1}{2n} \pi\right)}{\eta^2 + \sin^2\left(\frac{k}{n} \pi\right)} \quad (3)$$

With:  $\eta = \text{sh}\left(\frac{\text{Argch}(1/\varepsilon)}{n}\right)$

and  $\varepsilon = \sqrt{10^{\frac{A_{\max}}{10}} - 1} \quad (4)$

The values of normalized parameters  $g_i$  are given in table 1.

Table1:  $g_i$  parameters

$g_1$	$g_2$	$g_3$	$g_4$	$g_5$	$g_6$	$g_7$	$g_8$	$g_9$
1.19	1.45	2.15	1.6	2.22	1.6	2.15	1.44	1.19

### 3. Design of filter by LC elements

From the normalization parameters  $g_i$ , we deduce the values of inductances  $L_i$  and capacitance  $C_i$  by using the equations bellow:

$$L_i = \frac{g_i Z_0}{f_1} \quad (5)$$

$$C_i = \frac{g_i}{Z_0 f_1} \quad (6)$$

Table 2: Denormalized elements ( $L_i, C_i$ ) of filter

$g_i$ elements	Inductance $L_i$ (nH)	Capacitance $C_i$ (pF)
1.19	9.4745	
1.45		4.6178
2.15	17.11783	
1.6		5.09554
2.22	17.675	
1.6		5.09557
2.15	17.11783	
1.44		4.5859
1.19	9.4745	

We get the mounting of the filter given by Figure3

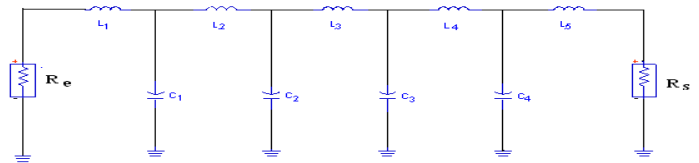


Fig. 4 LC low pass filter

The simulation of the above circuit by ADS software, gave us the characteristic of transmission and reflection coefficients as function as frequency below.

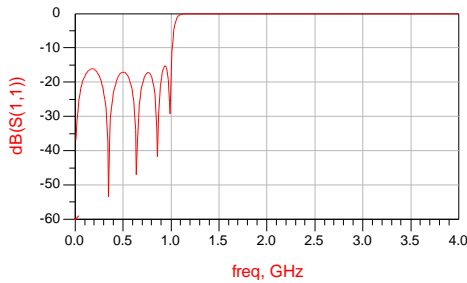


Fig. 5 Reflection coefficient S11 seen at entry.

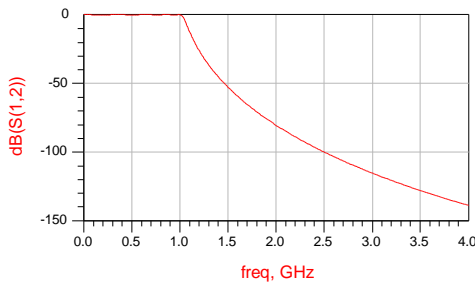


Fig. 6 Transmission coefficient S12 seen from exit to entry.

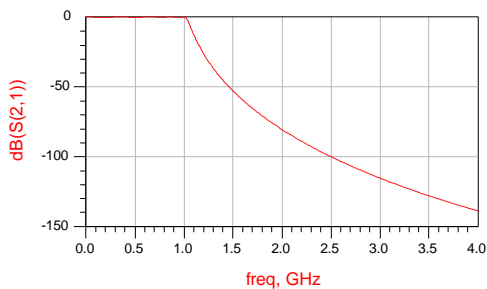


Fig. 7 Transmission coefficient S21 seen from entry to exit.

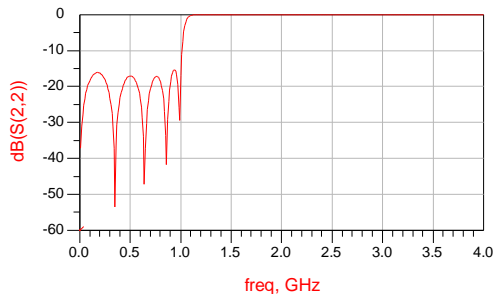


Fig. 8 Reflection coefficient S22 seen at exit.

The obtained results are in agreement with the ideal curve given by Figure 2.

#### 4. Design of filter by microstrip line

A microstrip line is a type of electrical transmission line [2]. It is used to convey microwave-frequency signals. We can design a high frequency filter by using cascaded microstrip lines with a number corresponding to the filter order [3].

In our study, the order of filter is 9; therefore, we need to 9 microstrip lines to conceive the filter.

The microstrip line that we will use is characterized by:

- A substrate thickness:  $h = 1.52\text{mm}$
- A permittivity:  $\epsilon_r = 4.32$
- A Copper Thickness:  $T = 0.035\text{mm}$
- A resistivity:  $\rho = 0.71$
- $Tang\Phi = 0.018$
- input impedance:  $Z_0 = 50\Omega$

We consider:

- The characteristic impedance of the capacitive line section:  $Z_{0C} = 10\Omega$
- The characteristic impedance of the inductive line section:  $Z_{0L} = 100\Omega$

By using the line calc of ADS software we calculated the width values of lines, we obtained  $W_C = 4\text{mm}$ , the width of capacitances and  $W_L = 0.8\text{mm}$ , the width of inductances.

The length values of lines  $l_{C_i}$  and  $l_{L_i}$  can be easily calculated from the values of the reactive elements and the characteristic impedances as follows:

$$l_{C_i} = C_i \cdot Z_{0C} \cdot \mathcal{G} \quad (7)$$

$$l_{L_i} = \frac{L_i \cdot \mathcal{G}}{Z_{0L}} \quad (8)$$

With  $\mathcal{G}$ : wave propagation velocity

The mounting of microstrip filter is given by Figure 9.



Fig. 9 The microstrip low pass filter

We obtained easily the layout of filter by ADS software.

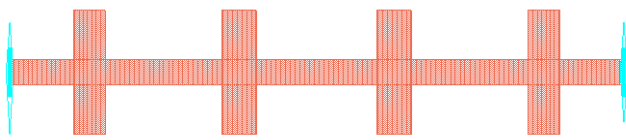


Fig.10 Layout of low pass filter

The simulation of the microstrip filter gave us the following results:

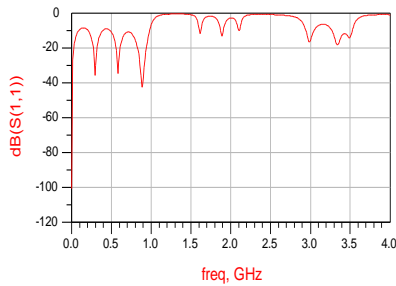


Fig.11 Reflection coefficient S11 seen at entry

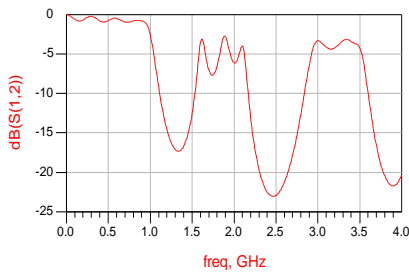


Fig.12 Transmission coefficient S12 seen from exit to entry

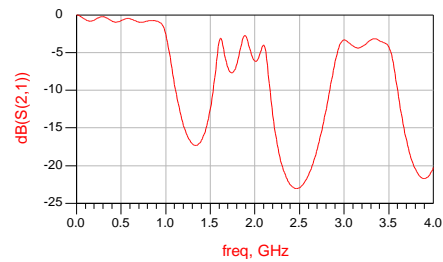


Fig. 13 Transmission coefficient S21 seen from entry to exit

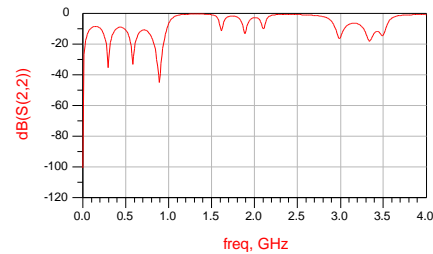


Fig. 14 Reflection coefficient S22 seen at exit

The microstrip filter presents the same cut-off frequency (1 GHz) than LC filter. However it presents parasitic responses that we can remove by changing the used uniform transmission lines by non-uniform Transmission Lines.

### 5. Scattering bond graph model of filter

Bond graph is a graphical representation of physical systems. This technique is based on exchange of energy and can give concise description of complex systems. By this approach, a physical system can be represented by symbols and lines, identifying the power flow paths. This method is widely used to model low frequency systems.

A new approach of analyzing high frequency system is developed by Mr Abdelkader Mami and Mr Hichem Taghouti in [4][5]. This new method combines the bond graph approach with the scattering formalism that used for several years to describe high frequency systems. This new technique is called scattering bond

graph method described clearly in [6][7]. It is used to give a concise model of filter and describes clearly the distribution of

energy between different elements of filter.

The bond graph model of the studied filter is given in Figure 15.

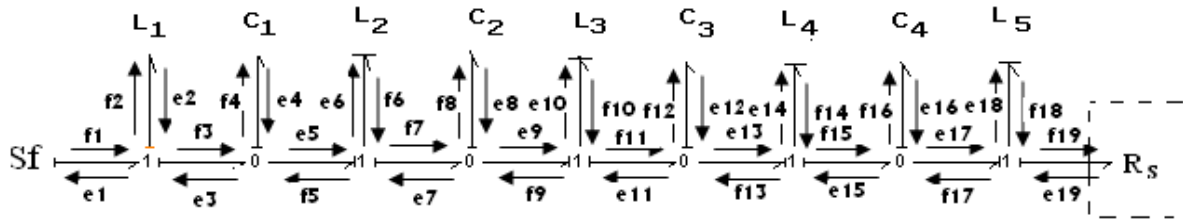


Fig. 15 Bond graph model of filter

e: effort  
 f: flow

To extract the scattering parameters from the bond graph representation and by using the new method which is described

in[4][5][6][7], we must transform the bond graph model given by figure15 into a causal bond graph model often named reduced bond graph given by Figure16.

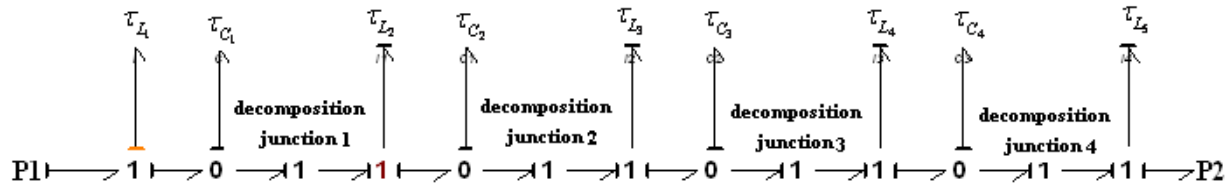


Fig. 16 The reduced bond graph method

$$\tau_{L_i} = \frac{L_i}{R_0} \quad (9)$$

By decomposition the reduced bond graph given by figure 16, we will have the following bond graph representation:

$$\tau_{C_i} = C_i \cdot R_0' \quad (10)$$

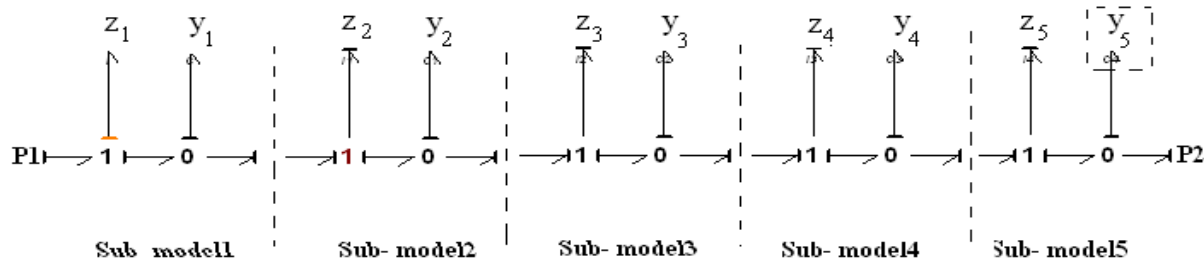


Fig. 17 Five bond graph sub model

$$z_i = \tau_{L_i} \cdot s \quad (11)$$

$$y_i = \tau_{C_i} \cdot s \quad (12)$$

We have the integro-differential operators by taking into account to the equations from the reduced bond graph model with effort-flow causality [7]:

$$B_i = -\frac{1}{z_i \cdot y_i} \quad (13)$$

Loop gain of sub-model i

$$\Delta_i = 1 + \frac{1}{z_i \cdot y_i} \quad (14)$$

Determinant of causal bond graph of the sub-model i

$$\begin{cases} H_{11} = \frac{z_i}{z_i \cdot y_i + 1} \\ H_{12} = \frac{1}{z_i \cdot y_i + 1} \\ H_{21} = \frac{1}{z_i \cdot y_i + 1} \\ H_{22} = -\frac{y_i}{z_i \cdot y_i + 1} \\ \Delta(s) = -\frac{1}{z_i \cdot y_i + 1} \end{cases} \quad (15)$$

The all integro-differential operators of the sub-model i. From these operators, we can deduce directly the wave matrix of each sub-model.

$$W^{(i)} = \frac{1}{2} \begin{bmatrix} z_i y_i - z_i - y_i + 2 & -z_i y_i + z_i + y_i \\ -z_i y_i - z_i + y_i & z_i y_i + z_i + y_i \end{bmatrix} \quad (16)$$

We deduced the wave matrix of the total model.

$$W^{(T)} = W^{(1)} * W^{(2)} * W^{(3)} * W^{(4)} * W^{(5)} = \begin{bmatrix} W_{11} & W_{12} \\ W_{21} & W_{22} \end{bmatrix} \quad (17)$$

From the total wave matrix, we deduced the scattering bond graph matrix S [5].

$$S = \begin{bmatrix} W_{22}^{-1} W_{12} & W_{11} & -W_{21} W_{12} W_{22}^{-1} \\ W_{22}^{-1} & & -W_{21} W_{22}^{-1} \end{bmatrix} \quad (18)$$

$$S_{11} = W_{22}^{-1} W_{12} \quad (19)$$

$$S_{12} = W_{11} - W_{21} W_{12} W_{22}^{-1} \quad (20)$$

$$S_{21} = W_{22}^{-1} \quad (21)$$

$$S_{22} = -W_{21} W_{22}^{-1} \quad (22)$$

From the determined equations of scattering parameters, we developed a maple program. Its simulation gave us the following results.

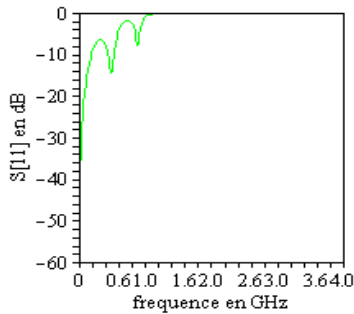


Fig. 18 Reflection coefficient S11 seen at entry

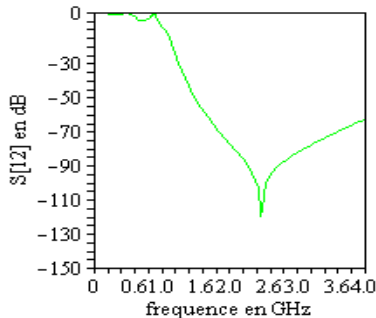


Fig. 19 Transmission coefficient S12 seen from exit to entry

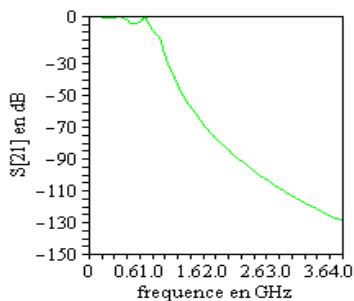


Fig. 20 Transmission coefficient S21 seen from entry to exit

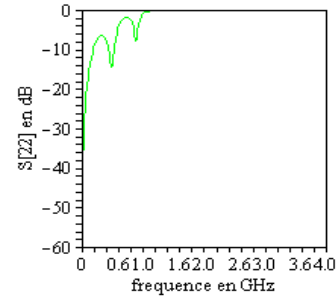


Fig. 21 Reflection coefficient S22 seen at exit

From the above results, we notice that the scattering bond graph method gives the same results of microstrip and RLC filter. So we confirm that the new method is useful to model high frequency systems.

## 6. Conclusion

In this paper, we presented a general study of high frequency filter. We developed the equations to conceive a high frequency low pass filter by reactive elements and by microstrip lines. And we used the scattering bond graph approach to give an explicit model of filter that describes the distribution of energy between its components.

We confirmed that the scattering bond graph method is appropriate to analyze and model high frequency systems.

## Acknowledgements

We would like to thank **Mr MAMI Abdelkader**, professor in Faculty of Sciences of Tunis (FST) and **Dr. TAGHOUTI Hichem**, Assistant Professor of Electrical Engineering in Superior institute of sciences and technologies of environment of BORJ-SEDRIA Tunis and researcher in Laboratory of Analysis and Command Systems, National Engineering School of Tunis for the time and guidance given, without forgetting the members of the unit of electronics and high frequency circuits specially and **Mr. MEHOUACHI Riadh**, doctoral researcher in Laboratory of Analysis and Command Systems, National Engineering School of Tunis and all those who contributed and aided for this study in particularly L.A.C.S members (Laboratory of analysis and command systems).

## References

- [1]. K.V.Puglia, "A general design procedure for bandpass filters derived from low pass prototype elements", Microwave journal 2000, 12
- [2]. Kin-Lu Wong, "Compact and Broadband Microstrip Antenna", Copyright 2002 John Wiley & Sons Inc. ISBNs: 0-471-41717-3 (Hardback); 0-471-22111-2 (Electronic)
- [3]. Benjamin Potelon, Etude et conception de filters hyperfréquences hybrids planaires- volumiques, these présentée à l'université de bretagne occidentale.
- [4]. Taghouti Hichem, Mami Abdelkader, "Application of the reduced bond graph approach to determinate the

- scattering parameters of a high frequency filter”, 10th International conference on sciences and techniques of automatic control & computer engineering December 20-22, 2009 Hammamet, Tunisia.
- [5]. Taghouti Hichem, Mami Abdelkader, “Extraction, Modelling and Simulation of the Scattering Matrix of a Chebychev Low-Pass Filter with cut-off frequency 100 MHz from its Causal and Decomposed Bond Graph Model”, ICGST-ACSE Journal, Volume 10, Issue 1, November 2010.
- [6]. Taghouti Hichem, Mami Abdelkader, “How to Find Wave-Scattering Parameters from the Causal Bond Graph Model of a High Frequency Filter”, American Journal of Applied Sciences 7 (5): 702-710, 2010, ISSN 1546-9239.
- [7]. Taghouti Hichem, Mami Abdelkader, “Modeling Method of a Low-Pass Filter Based on Microstrip T-Lines with Cut-Off Frequency 10 GHz by the Extraction of its Wave-Scattering”, Am. J. Eng. Appl. Sci., 3(4): 631-642.

**HICHEM TAGHOUTI:** Was born in Tunisia in December 1979. He received the master diploma in electronic (Numeric Analysis and Treatment of Electronics Systems) from the Faculty of Sciences of Tunis (FST) in 2005. Since September 2006, he has been an Assistant of Physics-Electronic in Faculty of Sciences of Gafsa in Tunisia. He was preparing his Thesis at National Engineering School of Tunis (ENIT) in Tunisia and faculty of Sciences of Tunis (FST) in Tunisia. His main research areas are the Bond Graph modelling and its New Applications in High Frequencies Domain. On April 2011 he received his PhD. actually he is working as Assistant Professor of Electrical Engineering in Superior institute of sciences and technologies of environment of BORJ-SEDRIA Tunis.



**SAMEH KHMAILIA:** Was born in Gafsa-Tunisia. She received the master diploma in electronic (Numeric Analysis and Treatment of Electronics Systems) from the Faculty of Sciences of Tunis (FST) in 2010. From February to September 2010, she was a contractual assistant of Physics-Electronic in Higher Institute of Accountancy and Entrepreneurial Administration of Mannouba in Tunisia. Since September 2010, she has been a contractual assistant of Physics-Electronic in Higher Institute of Applied Sciences and Technology of Gafsa in Tunisia. She prepares her Thesis in National Engineering School of Tunis (ENIT) in Tunisia and faculty of Sciences of Tunis (FST) in Tunisia. Her main research areas are the Scattering- Bond Graph modeling and its new Applications in High Frequencies Domain.



**RIADH MEHOUACHI:** Was born in SILIANA (Tunisia) in December 1978. He received the master diploma in Electronic (Numeric Analysis and Treatment of Electronics Systems) from the Faculty of Sciences of Tunis (FST) in 2007. Since September 2006, he has been a Technical teaching professor Tunis. Actually, he is preparing his thesis in National Engineering School of Tunis (ENIT) in Tunisia. His main research areas are the bond graph modeling and its new Applications in High Frequencies Domain.



**ABDELKADER MAMI:** Was born in Tunisia, he is a Professor in Faculty of Sciences of Tunis (FST). He was received his Dissertation H.D.R (Enabling To Direct of Research) from the University of Lille (France) 2003, he is a member of Advise Scientific in Faculty of Science of Tunis (Tunisia), he is a President of commutated thesis of electronics in the Faculty of sciences of Tunis, He is a person in charge for the research group of analyze and command systems in the ACS- laboratory in ENIT of Tunis and in many authors fields.

